

# Compressive Sensing Using Symmetric Alpha-Stable Distributions for Robust Sparse Signal Reconstruction

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**Abstract**—Traditional compressive sensing (CS) primarily assumes light-tailed models for the underlying signal and/or noise statistics. Nevertheless, this assumption is not met in the case of highly impulsive environments, where non-Gaussian infinite-variance processes arise for the signal and/or noise components. This drives the traditional sparse reconstruction methods to failure, since they are incapable of suppressing the effects of heavy-tailed sampling noise. The family of symmetric alpha-stable ( $S\alpha S$ ) distributions, as a powerful tool for modeling heavy-tailed behaviors, is adopted in this paper to design a robust algorithm for sparse signal reconstruction from linear random measurements corrupted by infinite-variance additive noise. Specifically, a novel greedy reconstruction method is developed, which achieves increased robustness to impulsive sampling noise by solving a minimum dispersion (MD) optimization problem based on fractional lower-order moments. The MD criterion emerges naturally in the case of additive sampling noise modeled by  $S\alpha S$  distributions, as an effective measure of the spread of reconstruction errors around zero, due to the lack of second-order moments. The experimental evaluation demonstrates the improved reconstruction performance of the proposed algorithm when compared against state-of-the-art CS techniques for a broad range of impulsive environments.

**Index Terms**—Compressive sensing, sparse recovery, symmetric alpha-stable distributions, heavy-tailed statistics, fractional lower-order moments, minimum dispersion criterion.

## I. INTRODUCTION

USING the concept of transform coding, compressive sensing (CS) enables a potentially large reduction in the sampling and computation costs for capturing signals that have

a *sparse* or *compressible* representation. Furthermore, CS is characterized by an intrinsic denoising mechanism, which suppresses the non-sparse contributions due to noise. This is extremely important in practical applications, where an observed signal and/or its corresponding set of compressive measurements are typically corrupted by noise.

In order to suppress the noise impact, which deteriorates the accurate reconstruction of the original signal from a reduced set of compressive measurements, a broad range of noise-aware sparse reconstruction algorithms have been developed. These include: greedy pursuit [1], [2], convex relaxation [3]–[5], Bayesian formulation [6], [7], nonconvex optimization [8]–[10] and brute force [11]. Each method has its own advantages and limitations. For instance, greedy pursuits and convex optimization are computationally more tractable and yield provably correct reconstructions under well-determined conditions. However, apart from sparsity, they are not able to account for any prior statistical information about the signal and noise, which could be used to improve the reconstruction accuracy. On the other hand, Bayesian methods and nonconvex optimization are typically based on rigorous principles, but often they do not provide theoretical guarantees. Finally, brute force approaches are algorithmically solid, but their practical use is restricted to small-scale problems.

The majority of previous CS reconstruction methods is primarily based on light-tailed, finite-variance assumptions for the statistics of the signal and/or noise generating processes. Despite the analytical tractability and practical appeal, these assumptions may yield a dramatic degradation of the reconstruction quality when we operate in highly impulsive environments, which give rise to *heavy-tailed* processes with infinite variance. To alleviate the effects of gross errors that mask the information conveyed by the compressive measurements, recent state-of-the-art methods rely on algebraic-tailed models, specifically on Cauchy and generalized Cauchy (GCD) distributions, to design robust reconstruction methods [12], [13].

On the other hand, *alpha-stable* distributions [14] have been proven very powerful in accurately modeling impulsive phenomena. However, their intractability due to the lack of closed-form expressions for the density functions of all except for a few stable distributions (Gaussian, Cauchy and Lévy) has prevented their exploitation in the framework of CS. To address this problem, whilst also revealing the advantages of alpha-stable models in designing efficient CS systems, this paper

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proposes a method for robust sparse signal reconstruction under heavy-tailed sampling noise, by modeling the noise statistics via *symmetric alpha-stable* (S $\alpha$ S) distributions. To the best of our knowledge, this is the first thorough study that bridges the fields of sparse signal reconstruction and S $\alpha$ S models for the design of a CS system with increased robustness to heavy-tailed infinite-variance sampling noise.

### A. Motivation

In practical CS acquisition systems, the generated compressive measurements are typically corrupted by *sampling* noise. The presence of large-amplitude noise in the measurement domain degrades dramatically the reconstruction accuracy of traditional CS techniques based on  $\ell_1$  or  $\ell_2$  norms. The problem becomes even more challenging when we operate in impulsive environments, where the corrupting sampling noise can be of infinite variance. This causes conventional sparse reconstruction algorithms to fail in recovering a close approximation of the original signal.

The problem of accurate sparse signal reconstruction from random measurements corrupted by gross sampling errors has been previously addressed in the context of error correction coding [16] and incomplete measurements [17], [18]. The main limitation of these approaches is that their reconstruction performance relies on the sparsity of the error term, which is a condition that may not be met often in practice.

Recently, the problem of robust sparse signal reconstruction from random measurements corrupted by impulsive sampling noise has been addressed efficiently by employing the Lorentzian norm either as an objective function or as a constraint. In particular, in [12], [19] the true sparse signal is reconstructed by solving an  $\ell_0$ -regularized least logarithmic deviation problem. The logarithmic deviation is defined in terms of the Lorentzian norm, which does not over-penalize large deviations, and is therefore more robust than the commonly used  $\ell_1$  and  $\ell_2$  norms for the suppression of impulsive noise. A similar approach is proposed in [13], where a nonconvex optimization problem is solved to reconstruct the sparse signal by minimizing the  $\ell_1$  norm subject to a nonlinear constraint based on the Lorentzian norm. The use of the Lorentzian norm in these papers is further justified by the existence of logarithmic moments for heavy-tailed distributions, whose second-order moments are infinite or even undefined.

Despite the enhanced reconstruction quality of the above methods in the presence of heavy-tailed additive sampling noise, the specific use of Cauchy [12], [19] or GCD distributions [13] can be restrictive in capturing more generic non-Gaussian heavy-tailed behaviors of the sampling noise. Motivated by this limitation, first we model the statistics of highly impulsive sampling noise, with possibly infinite variance, by members of the S $\alpha$ S family. Then, we propose a novel greedy sparse reconstruction algorithm, which minimizes the *dispersion* of the random measurements' error. As it will become clear in the subsequent analysis, the minimum dispersion (MD) criterion arises naturally as a measure of the spread of estima-

tion errors around zero for random variables modeled by S $\alpha$ S distributions, as is the case with the infinite variance sampling noise adopted in this study. Most importantly, we show that the MD criterion is equivalent to a minimum  $\ell_p$  estimation error criterion, which simplifies the design of our proposed sparse reconstruction algorithm. Nevertheless, we emphasize that the subsequent analysis considers a conventional linear sampling operator for the generation of random measurements.

### B. Main Contributions

The major contribution of this paper is twofold: i) we propose a novel iterative greedy algorithm that combines the characteristics of a gradient-descent approach with a statistical optimization criterion, namely, the *minimum dispersion* (MD) criterion, which is equivalent to minimizing the *fractional lower-order moments* (FLOMs) of reconstruction errors. The FLOMs measure the  $\ell_p$  ( $p < 2$ ) distance between the reconstructed and the true sparse signal; ii) we provide theoretical guarantees for the convergence of the algorithm and an upper bound of the  $\ell_2$  norm of the reconstruction error, along with rules of thumb for setting the key parameters that control the performance of the proposed algorithm.

Furthermore, the use of  $\ell_p$  distance metrics with  $p < 2$ , which arise naturally when S $\alpha$ S models are coupled with a minimum dispersion rule, provides an additional degree of freedom (i.e., the value of  $p$ ) yielding increased robustness against gross sampling errors. The proposed reconstruction algorithm resembles an orthogonal matching pursuit (OMP) approach in the sense that, at each step, it selects the measurement basis vector which is most correlated with the current residuals. However, the key difference between our algorithm and an OMP-based approach is that this correlation is expressed in terms of FLOMs, thus it adapts perfectly to non-Gaussian, heavy-tailed processes with infinite variance. At each iteration, one or several elements of the sparse signal are reconstructed, therefore, as the algorithm progresses, a refined estimate of its nonzero elements is obtained by removing the contribution of previously estimated elements.

### C. Paper Organization

The rest of the paper is organized as follows: Section II introduces the minimum dispersion as a proper optimization criterion for heavy-tailed infinite-variance sampling noise modeled by S $\alpha$ S distributions, and proves its equivalence with a minimum  $\ell_p$  estimation error criterion. Section III analyzes the design and implementation of our proposed iterative greedy algorithm for sparse signal reconstruction, whilst also providing theoretical proofs for its convergence, along with an upper bound of the reconstruction error. An experimental evaluation of the reconstruction performance is presented in Section IV for a variety of impulsive environments, where our proposed algorithm is compared against state-of-the-art sparse reconstruction methods tailored to impulsive sampling noise. Finally, Section V concludes and gives directions for future work.

#### D. Notation

In the following, scalars are denoted by lower-case letters (e.g.  $x$ ), column vectors by lower-case boldface letters (e.g.  $\mathbf{x}$ ), and matrices by upper-case boldface letters (e.g.  $\mathbf{X}$ ). Sets are represented by calligraphic letters (e.g.  $\mathcal{S}$ ), while  $|\mathcal{S}|$  denotes their cardinality. The  $i$ th column of a matrix  $\mathbf{X}$  is denoted by  $\mathbf{x}_i$ , whereas  $x_j$  indicates the  $j$ th element of a vector  $\mathbf{x}$ .  $\mathcal{S}_i$  denotes a subset of  $\mathcal{S}$ , while  $\mathbf{X}_{\mathcal{S}}$  is the submatrix formed by the columns  $\{\mathbf{x}_i \mid i \in \mathcal{S}\}$ , whose indices belong to  $\mathcal{S}$ . Similarly,  $\mathbf{x}_{\mathcal{S}}$  is the subvector formed by the elements  $\{x_j \mid j \in \mathcal{S}\}$ , whose indices belong to  $\mathcal{S}$ . Finally, we use  $\hat{\mathbf{x}}$ ,  $\mathbf{x}^T$ ,  $\mathbf{x}^*$ , and  $\mathbf{x}^{(t)}$  to denote the estimate (reconstruction), transpose, optimal solution, and value at  $t$ th iteration of a vector  $\mathbf{x}$ , respectively. Similar notations are used for the matrices.

## II. SPARSE RECONSTRUCTION VIA A MINIMUM DISPERSION CRITERION

Let  $\mathbf{x} = [x_1, x_2, \dots, x_N]^T \in \mathbb{R}^N$  be a real discrete-time signal. In the general case, we assume that  $\mathbf{x}$  can be sparsified over a, possibly overcomplete, transform basis  $\Psi \in \mathbb{R}^{N' \times N}$  with  $N' \geq N$ , such that  $\alpha = \Psi \mathbf{x} \in \mathbb{R}^{N'}$  is an  $s$ -sparse vector of transform coefficients.  $\Psi$  and  $\Psi^T$  denote the analysis (direct) and synthesis (inverse) transforms, respectively. In the following, a linear sampling operator is employed for the generation of a reduced set of compressive measurements.

Given a random measurement matrix  $\Phi \in \mathbb{R}^{M \times N}$  ( $M < N$ ), which satisfies all the necessary and sufficient conditions for accurate sparse reconstruction, the generic noisy sampling model adopted in the subsequent analysis is as follows,

$$\mathbf{y} = \Phi \Psi^T \alpha_0 + \mathbf{n}, \quad (1)$$

where  $\alpha_0 \in \mathbb{R}^{N'}$  is the true sparse signal to be recovered,  $\mathbf{y} \in \mathbb{R}^M$  is a vector of  $M$  noisy random measurements, and  $\mathbf{n} \in \mathbb{R}^M$  is the additive sampling noise. By setting  $\mathbf{A} = \Phi \Psi^T$ , various well-established approaches for recovering the sparse signal solve a constrained optimization problem of the form,

$$\min_{\alpha \in \mathbb{R}^{N'}} \|\mathbf{y} - \mathbf{A}\alpha\|_p \quad \text{s.t.} \quad \|\alpha\|_q \leq s, \quad (2)$$

or a regularized optimization problem,

$$\min_{\alpha \in \mathbb{R}^{N'}} (\|\mathbf{y} - \mathbf{A}\alpha\|_p + \tau \|\alpha\|_q), \quad (3)$$

with  $0 < q \leq 1$ ,  $p \in \{2, \infty\}$  or denoting the Lorentzian norm,<sup>1</sup>  $s$  being the sparsity level and  $\tau$  a regularization parameter that balances the influence of the data fidelity term and the sparsity-inducing term on the optimal solution. Having obtained an estimate of the optimal sparse coefficients vector  $\alpha_0^*$ , the original signal is given by  $\hat{\mathbf{x}}_0 = \Psi^T \alpha_0^*$ .

At the core of our proposed sparse reconstruction algorithm is the use of S $\alpha$ S distributions for modeling the statistics of impulsive sampling noise. In particular, we assume that the random noise  $\mathbf{n}$  consists of independent and identically distributed

(i.i.d.) components,  $n_i \in \mathbb{R}$ ,  $i = 1, \dots, M$ , that follow a univariate S $\alpha$ S distribution. The probability density function of a general univariate S $\alpha$ S distribution is as follows [15],

$$f_\alpha(x; \gamma, \delta) = \frac{1}{\gamma} h\left(\frac{x - \delta}{\gamma}; \alpha\right), \quad (4)$$

where

$$h(x; \alpha) = \frac{1}{\pi} \int_0^\infty \cos(xt) e^{-t^\alpha} dt. \quad (5)$$

In the above expressions,  $\alpha \in (0, 2]$  is the *characteristic exponent*,  $\gamma > 0$  is the *dispersion*, and  $\delta \in \mathbb{R}$  is the *location parameter* of the distribution. The characteristic exponent is a shape parameter, which controls the thickness of the tails of the density function. The smaller the  $\alpha$ , the heavier the tails of the S $\alpha$ S density function. The dispersion parameter determines the spread of the distribution around its location, as the standard deviation does for a Gaussian distribution. In most practical applications, the estimated  $\alpha$  is typically larger than or equal to 1. Furthermore, the case of  $\alpha = 2$  corresponds to the Gaussian, whereas our focus is on non-Gaussian heavy-tailed statistics for the sampling noise. Because of these remarks, we emphasize that, without loss of generality, our subsequent analysis considers for the distribution of the noise components that  $n_i \sim f_{\alpha_n}(\gamma_n, \delta_n)$  with  $\alpha_n \in [1, 2)$  and  $\delta_n = 0$ . In the following, the subscript  $n$  is used to denote variables related with the additive sampling noise  $\mathbf{n}$ .

Due to their algebraic tails, S $\alpha$ S distributions lack finite second-order moments. Instead, all moments of order  $p$  less than  $\alpha$  do exist and are called the *fractional lower-order moments* (FLOMs). In particular, the FLOMs of a S $\alpha$ S random variable  $X \sim f_\alpha(\gamma_X, 0)$  are given by [14]

$$\mathbb{E}\{|X|^p\} = (C_{p,\alpha} \gamma_X)^p, \quad 0 < p < \alpha, \quad (6)$$

where

$$(C_{p,\alpha})^p = \frac{2^{p+1} \Gamma(\frac{p+1}{2}) \Gamma(-\frac{p}{\alpha})}{\alpha \sqrt{\pi} \Gamma(-\frac{p}{2})} = \frac{\Gamma(1 - \frac{p}{\alpha})}{\cos(\frac{\pi}{2}p) \Gamma(1-p)}. \quad (7)$$

In addition, (6) yields the following expression for the dispersion of  $X$  in terms of the FLOMs,

$$\gamma_X = (\mathbb{E}\{|X|^p\})^{1/p} C_{p,\alpha}^{-1}, \quad (8)$$

which will be employed to quantify the spread of gross noise samples around zero.

Furthermore, due to the lack of finite variance for alpha-stable distributed data, the traditional minimum mean squared error (MMSE) criterion cannot be used as a measure of the reconstruction quality to be optimized. Instead, we employ the *minimum dispersion* (MD) criterion in our optimization problem, since the dispersion of alpha-stable random variables is finite and gives a good measure of the spread of estimation errors around zero. Most importantly, our proposed reconstruction method belongs to the class of  $\ell_p$ -based nonlinear, nonconvex relaxation techniques. Specifically,  $\ell_p$  (quasi-)norms ( $0 < p < 2$ ), and subsequently the use of a *least  $\ell_p$  estimation error* criterion, emerge naturally in the case of infinite variance sampling noise modeled as a S $\alpha$ S random variable. Indeed, as we deduce

<sup>1</sup>The  $\ell_p$  norm of a vector is defined by  $\|\mathbf{x}\|_p = (\sum_{j=1}^N |x_j|^p)^{1/p}$  for  $0 < p \leq 2$  ( $\ell_p$  is a quasi-norm for  $0 < p < 1$ ), whilst the Lorentzian norm is defined by  $\|\mathbf{x}\|_{LL_2} = \sum_{j=1}^N \log(1 + x_j^2)$ .

TABLE I  
OPTIMAL  $p$  PARAMETER AS A FUNCTION OF THE CHARACTERISTIC  
EXPONENT  $\alpha$

$\alpha$	1	1.1	1.2	1.3	1.4	1.5	1.6	1.7	1.8	1.9	2
$p_{\text{opt}}$	0.52	0.56	0.58	0.61	0.64	0.69	0.72	0.76	0.81	0.88	0.98

from (8), an  $\ell_p$  (quasi-)norm based approximation of the dispersion of  $X$  can be obtained by replacing the FLOM,  $\mathbb{E}\{|X|^p\}$ , with a discrete finite sum,

$$\mathbb{E}\{|X|^p\} \approx \frac{1}{N} \sum_{j=1}^N |x_j|^p = \frac{1}{N} \|\mathbf{x}\|_p^p. \quad (9)$$

The larger the sample size, the more accurate this approximation will be.

Let  $\hat{X}$  denote an estimate of the random variable  $X$  and  $E = X - \hat{X}$  be the estimation error. Then, the minimum dispersion criterion can be viewed as a minimum  $\ell_p$  estimation error criterion. Indeed, from (8) and (9) we deduce that minimizing the dispersion of the error,  $\gamma_E$ , is equivalent to minimizing the  $\ell_p$  (quasi-)norm of the associated error vector  $\mathbf{e} \in \mathbb{R}^N$ , consisting of  $N$  realizations of  $E$ .

#### A. Estimation of $p$

Notice that the selection of an appropriate value for the  $p$  parameter in the above expressions is a critical step. Most importantly, the optimal  $p$  depends on the characteristic exponent  $\alpha = \alpha_n$ , which is estimated from the noisy compressive measurements. A method for choosing the optimal  $p$  as a function of  $\alpha$  has been proposed in [20], which is based on minimizing the standard deviation of a FLOM-based covariation estimator.<sup>2</sup> Other authors suggest the optimal  $p$  should be lower but as close as possible to the value of  $\alpha$ , if  $\alpha$  can be inferred. However, this approach causes the  $p$ th FLOM to approach infinity as  $p \rightarrow \alpha$ , since  $C_{p,\alpha}$  in (6) goes to infinity. On the other hand, the FLOM-based approach yields an almost linear relation between  $\alpha$  and the optimal value of  $p$ , and specifically  $p \lesssim \alpha/2$ . In addition, if  $p < \alpha/2$  the FLOM estimator has a finite variance, which is desirable [21]. In the following, we adopt this rule and the optimal value of  $p$  is set as a function of  $\alpha$  by linearly interpolating the entries of the lookup Table I, generated by the FLOM-based approach in [20].

### III. ROBUST SPARSE RECONSTRUCTION ALGORITHM

In this section, we propose a novel iterative greedy reconstruction algorithm, which suppresses efficiently the effects of heavy-tailed sampling noise of infinite variance, whilst achieving increased robustness to a broader range of impulsive noise behaviors, from near linear (i.e.,  $\alpha \rightarrow 2$ ) to extremely impulsive sampling noise (i.e.,  $\alpha \rightarrow 1$ ). To this end, we analyze the design and implementation of our proposed algorithm, which solves an  $\ell_0$ -constrained  $\ell_p^p$  optimization problem. For convenience, yet

<sup>2</sup>The FLOM-based covariation estimator for two jointly  $S\alpha S$  random variables  $X, Y$  with  $\alpha > 1$ , is defined by  $c_{XY} = \frac{\sum_{i=1}^N x_i |y_i|^{p-1} \text{sign}(y_i)}{\sum_{i=1}^N |y_i|^p} \gamma_Y^\alpha$ .

without loss of generality, in the following theoretical derivations we assume that the original signal  $\mathbf{x}_0$  is sparse in the canonical basis  $\Psi = \mathbf{I}$ , thus  $\mathbf{x}_0 = \alpha \mathbf{0}$ .

#### A. $\ell_0$ -Constrained Dispersion Minimization

In the following, we aim at designing a robust sparse reconstruction algorithm, expressed as an operator  $R : \mathbb{R}^M \mapsto \mathbb{R}^N$ , which reconstructs accurately the true sparse signal  $\mathbf{x}_0$  from a highly reduced set of noisy linear measurements  $\mathbf{y}$ . This algorithm must be robust, especially when we operate in highly impulsive environments, in the sense that small perturbations to the noiseless measurements should yield small perturbations in the reconstructed signal, even when a portion of the measurements is corrupted by large-amplitude noise.

To this end, the presence of the free parameter  $p$  in our proposed  $\ell_p$ -based penalization of the residual increases the robustness against gross outliers, as opposed to the recently used Lorentzian norm. This is because, in contrast to the Lorentzian that is intrinsically related with a Cauchy model (fixed  $\alpha = 1$ ), in our method the value of  $p$  depends on the inherent impulsiveness of the noise as expressed by its estimated characteristic exponent  $\alpha$ . It is important to emphasize that, given the constraints of Section II, namely, i)  $0 < p < \alpha$ , for the existence of FLOMs, ii)  $\alpha \in [1, 2)$ , and iii)  $p < \alpha/2$  for the selection of  $p$  as a function of  $\alpha$ , we obtain that  $0 < p < 1$ . This turns  $\ell_p$  into a quasi-norm, which has the tradeoff of better approximating the ideal  $\ell_0$  objective function for sparse reconstruction, but with an increased computational complexity due to its highly nonconvex nature.

More specifically, from (8) and (9) the original sparse signal  $\mathbf{x}_0$  is recovered by minimizing the dispersion of the data fidelity term, constrained on the maximum number of nonzero elements of  $\mathbf{x}_0$ , that is,

$$\min_{\mathbf{x} \in \mathbb{R}^N} \left( \frac{C_{p,\alpha}^{-p}}{M} \|\mathbf{y} - \Phi \mathbf{x}\|_p^p \right) \text{ s.t. } \|\mathbf{x}\|_0 \leq s. \quad (10)$$

By noticing that, for the feasible ranges of  $\alpha$  and  $p$  considered herein, namely,  $1 \leq \alpha < 2$  and  $0 < p < 1$ , respectively, the constant  $\frac{1}{M} C_{p,\alpha}^{-p}$  is always positive (see Appendix), then, the original sparse signal  $\mathbf{x}_0$  is reconstructed by solving an  $\ell_0$ -constrained least  $\ell_p^p$  optimization problem,

$$\min_{\mathbf{x} \in \mathbb{R}^N} \|\mathbf{y} - \Phi \mathbf{x}\|_p^p \text{ s.t. } \|\mathbf{x}\|_0 \leq s. \quad (11)$$

Nevertheless, the numerical solution of (11) can be extremely complex, even for moderate signal sizes. Therefore, the design of a computationally tractable algorithm is imperative to solve the above nonconvex, combinatorial, sparse recovery problem. To this end, in [22], [23] it was shown that, depending on the restricted isometry constant of the measurement matrix  $\Phi$  and the noise level, the resulting  $\ell_p^p$  optimization problem provides better theoretical guarantees in terms of stability and robustness than  $\ell_1$  minimization, whereas a *local minimizer* can be computed in polynomial time. In general, it is difficult to compute a global minimizer of a nonconvex functional. On the other hand, any local optimization method can perform well if initialized by a point sufficiently close to the global optimum. In the

following, an iterative algorithm is proposed for solving (11), along with some remarks on its proper initialization.

### B. Gradient Projection Iterative Hard Thresholding

In this section, we derive a suboptimal approach to solve (11) by employing a gradient projection (GP) formulation combined with an iterative hard thresholding (IHT) algorithm. This approach does not require the computationally intense process of matrix inversion, while providing near-optimal error guarantees [24].

More specifically, given the measurements vector  $\mathbf{y}$  and the measurement matrix  $\Phi$ , let  $\mathbf{x}^{(t)}$  denote the estimated sparse solution at  $t$ th iteration. Note that in the general case, where the original signal is not sparse by itself but can be sparsified in an appropriate transform domain, the measurement matrix is replaced by  $\mathbf{A} = \Phi\Psi^T$  and the solution corresponds to a sparse coefficients vector  $\alpha_0^*$  (ref. Section II).

Even though the optimization problem (11) is nonconvex and NP-hard (ref. [23]), there are several advantages of using this approach. First, according to [22], for a Gaussian measurement matrix  $\Phi$ , the restricted  $p$ -isometry property of order  $s$  holds if  $s$  is almost proportional to  $M$  when  $p \rightarrow 0^+$ . Second, in [8], [25] it is demonstrated that when  $\delta_{2s} < 1$ , the solution of an  $\ell_p$  minimization problem is sparse when  $p > 0$  is small enough, where  $\delta_{2s}$  is the restricted isometry constant (RIC) of  $\Phi$ . The third advantage is that the  $\ell_p$  minimization can be applied to a broader class of measurement matrices, for instance, when  $\Phi$  is a random matrix whose entries are i.i.d. pre-Gaussian random variables [26].

Due to the singularity of the gradient of the objective function in (11) because of the sparsity of the solution, we employ the following regularized version of the  $\ell_p^p$  introduced in [27],

$$\|\mathbf{x}\|_{p,\epsilon}^p = \sum_{j=1}^N (|x_j|^2 + \epsilon)^{p/2}, \quad (12)$$

where  $\epsilon > 0$  is a parameter that will go to zero in order to approximate  $\|\mathbf{x}\|_p^p$ . By combining (11) and (12), our  $\ell_0$ -constrained  $\ell_p^p$  optimization problem becomes as follows,

$$\min_{\mathbf{x} \in \mathbb{R}^N} \|\mathbf{y} - \Phi\mathbf{x}\|_{p,\epsilon}^p \text{ s.t. } \|\mathbf{x}\|_0 \leq s. \quad (13)$$

The above problem has a solution for any  $p \in (0, 1)$  and  $\epsilon > 0$ , where  $\epsilon$  can be also seen as an indicator of whether an element of the local minimizer is zero or not. We will derive an iterative algorithm to compute a solution of (13), proving that the algorithm converges for any starting point.

More specifically, following an iterative hard thresholding approach, at each iteration the algorithm computes the updated solution as follows,

$$\mathbf{x}^{(t+1)} = \mathbf{H}_s \left( \mathbf{x}^{(t)} + \mu^{(t)} \mathbf{g}^{(t)} \right), \quad (14)$$

where  $\mathbf{H}_s(\mathbf{x})$  denotes the hard thresholding operator, which sets all except for the largest (in magnitude)  $s$  elements of  $\mathbf{x}$  to zero,  $\mu^{(t)}$  is a step size and  $\mathbf{g}^{(t)}$  is a search direction. In our implementation we adopt a gradient projection scheme, thus the search direction is determined by the gradient of the objective

function in (13). After some algebraic manipulation, the gradient vector is given by

$$\mathbf{g}^{(t)} = \nabla_{\mathbf{x}} \|\mathbf{y} - \Phi\mathbf{x}^{(t)}\|_{p,\epsilon}^p = -p \Phi^T \mathbf{W}^{(t)} (\mathbf{y} - \Phi\mathbf{x}^{(t)}), \quad (15)$$

where  $\mathbf{W}^{(t)}$  is a  $M \times M$  diagonal matrix, whose main diagonal elements are as follows,

$$\mathbf{W}_{i,i}^{(t)} = \frac{1}{\left( (y_i - \phi_i^T \mathbf{x}^{(t)})^2 + \epsilon \right)^{1-\frac{p}{2}}}, \quad i = 1, \dots, M. \quad (16)$$

Notice also that the above weight matrix resembles the one computed when working in an iteratively reweighted least squares (IRLS) framework [28]. Furthermore, the inherent role of the weights is to suppress the effect of large errors by assigning a small weight when large deviations are estimated. In the special case of  $\mathbf{W}^{(t)} = \mathbf{I}$  the above GP-based IHT method reduces to the conventional least squares IHT [24]. Moreover, the hard thresholding operation may not yield a unique output. In this case, we select the  $s$  elements either at random or based on a predetermined ordering. In the rest of the text, we will refer to our proposed algorithm for minimizing the error dispersion using the above gradient projection hard thresholding technique as MD-IHT.

### C. Key Parameters Setting and Convergence Analysis

The efficiency of our proposed MD-IHT algorithm is affected by the accurate tuning of two key parameters, namely, the parameter  $p$ , which determines the  $\ell_p^p$  objective function and depends on the noise characteristic exponent, and the step size  $\mu^{(t)}$ .

Concerning the value of  $p$ , as we have already mentioned in Section II, we adopt the almost linear relation  $p \lesssim \alpha/2$ . Thus the problem is reduced to estimating accurately the noise characteristic exponent from the random measurements  $\mathbf{y}$ . Following the approach suggested by [28], the method of log-cumulants is employed to estimate the S $\alpha$ S parameters by equating sample log-cumulants to their theoretical counterparts for a particular model and then solving the resulting system, much in the same way as in the classical method of moments. In particular, by applying the Mellin transform on a S $\alpha$ S density we get the following expression for its second-kind first characteristic function,

$$\Phi(z) = \frac{\gamma^{z-1} 2^z \Gamma\left(\frac{z}{2}\right) \Gamma\left(-\frac{z-1}{\alpha}\right)}{\alpha \sqrt{\pi} \Gamma\left(-\frac{z-1}{2}\right)}. \quad (17)$$

Notice that by setting  $z = p + 1$  in  $\Phi(z)$  we obtain the expression of the FLOMs of a S $\alpha$ S random variable, as defined in (6). By taking the limit as  $z \rightarrow 1$  of the first and second derivatives of the  $\log(\Phi(z))$ , we derive the following results for the second-kind cumulants of a S $\alpha$ S model,

$$\tilde{k}_1 = \frac{\alpha - 1}{\alpha} \psi(1) + \log(\gamma), \quad \tilde{k}_2 = \frac{\pi^2}{12} \frac{\alpha^2 + 2}{\alpha^2}, \quad (18)$$

where  $\psi(\cdot)$  is the Digamma function. On the other hand, the first two sample second-kind cumulants can be estimated empirically

from the  $M$  measurements  $\mathbf{y}$  as follows,

$$\hat{k}_1 = \frac{1}{M} \sum_{i=1}^M \log(|y_i|), \quad \hat{k}_2 = \frac{1}{M} \sum_{i=1}^M \left( \log(|y_i|) - \hat{k}_1 \right)^2. \quad (19)$$

The estimation process simply involves solving (18) for  $\alpha$  and  $\gamma$  by substituting  $\tilde{k}_1, \tilde{k}_2$  with their sample estimates  $\hat{k}_1, \hat{k}_2$ , respectively.

Regarding the tuning of the step size, we notice that the convergence performance of the MD-IHT algorithm improves if an adaptive step size,  $\mu^{(t)}$ , is employed to normalize the gradient update in (14). Specifically, let  $\mathcal{S}^{(t)}$  be the support of  $\mathbf{x}^{(t)}$ , and assume that the algorithm has identified the true support of  $\mathbf{x}_0$ , that is,  $\mathcal{S}^{(t+1)} = \mathcal{S}^{(t)} = \mathcal{S}$ . Then, we want to minimize  $\|\mathbf{y} - \Phi_{\mathcal{S}} \mathbf{x}_S\|_{p,\epsilon}^p$  using a gradient projection algorithm with updates

$$\mathbf{x}_S^{(t+1)} = \mathbf{x}_S^{(t)} + \mu^{(t)} \mathbf{g}_S^{(t)}. \quad (20)$$

Optimality in estimating  $\mu^{(t)}$  is equivalent to finding a step size which reduces maximally the  $\ell_{p,\epsilon}^p$  objective function at each iteration. This is a nontrivial task and, in general, there is no known closed form for an optimal step size. To address this issue, we update the step size at each iteration in a suboptimal way according to the following proposition:

*Proposition 1:* A suboptimal adaptive step size  $\mu^{(t)}$ , which guarantees that the  $\ell_{p,\epsilon}^p$  objective function in (11) does not increase at each iteration, is given by

$$\begin{aligned} \mu^{(t)} &= \arg \min_{\mu} \left\| \mathbf{y} - \Phi_{\mathcal{S}} \left( \mathbf{x}_S^{(t)} + \mu \mathbf{g}_S^{(t)} \right) \right\|_{p,\epsilon}^p \\ &= \arg \min_{\mu} \left\| \left( \mathbf{y} - \Phi_{\mathcal{S}} \mathbf{x}_S^{(t)} \right) - \mu \Phi_{\mathcal{S}} \mathbf{g}_S^{(t)} \right\|_{p,\epsilon}^p. \end{aligned} \quad (21)$$

*Proof:* By setting  $\mathbf{u} = (\mathbf{y} - \Phi_{\mathcal{S}} \mathbf{x}_S^{(t)})$  and  $\mathbf{v} = \Phi_{\mathcal{S}} \mathbf{g}_S^{(t)}$ , we deduce that the estimation of a suboptimal step size  $\mu^{(t)}$  at the  $t$ th iteration is reduced to solving the following optimization problem,

$$\mu^{(t)} = \arg \min_{\mu} J_{p,\epsilon}(\mu) \triangleq \arg \min_{\mu} \|\mathbf{u} - \mu \mathbf{v}\|_{p,\epsilon}^p, \quad (22)$$

whose treatment depends on the value of  $p$ . We emphasize that in our proposed algorithm the value of  $p$  varies in the interval  $(0, 1)$ , thus the solution of (22) is examined only for this specific case.

First, we observe that, although for  $p \in (0, 1)$  the minimization of  $J_{p,\epsilon}(\mu)$  corresponds to a nonconvex problem with several local minima, however, it is still feasible to find a global minimizer. In particular, we start by defining the fractions  $\{r_i = \frac{u_i}{v_i}\}_{i=1}^M$ , which are then sorted in ascending order,  $r_{(1)}, \dots, r_{(M)}$ . Then, the optimization problem in (22) can be reformulated as follows,

$$\mu^{(t)} = \arg \min_{\mu} J_{p,\epsilon}^{\text{ord}}(\mu) \triangleq \arg \min_{\mu} \sum_{i=1}^M \left( |v_{[i]}|^2 |\mu - r_{(i)}|^2 + \epsilon \right)^{\frac{p}{2}} \quad (23)$$

where  $\{|v_{[i]}|\}_{i=1}^M$  denote the corresponding concomitant weights. Doing so, the domain of the objective function

$J_{p,\epsilon}^{\text{ord}}(\mu)$  is the union of  $M + 1$  adjacent intervals, namely,  $(-\infty, r_{(1)}], (r_{(1)}, r_{(2)}], \dots, (r_{(M-1)}, r_{(M)}], (r_{(M)}, +\infty)$ . Let  $\mu \in (r_{(m-1)}, r_{(m)})$  for some  $1 \leq m \leq M$ , then, the objective function becomes

$$\begin{aligned} J_{p,\epsilon}^{\text{ord}}(\mu) &= \sum_{i=1}^{m-1} |v_{[i]}|^p \left( (\mu - r_{(i)})^2 + \frac{\epsilon}{|v_{[i]}|^2} \right)^{\frac{p}{2}} \\ &\quad + \sum_{i=m}^M |v_{[i]}|^p \left( (r_{(i)} - \mu)^2 + \frac{\epsilon}{|v_{[i]}|^2} \right)^{\frac{p}{2}}. \end{aligned} \quad (24)$$

We notice that  $J_{p,\epsilon}^{\text{ord}}(\mu)$  is concave as a nonnegative combination of concave functions. Specifically,  $J_{p,\epsilon}^{\text{ord}}(\mu)$  is piecewise concave in each interval, thus it attains its local minima among the boundary points  $\{r_{(i)}\}_{i=1}^M$ . Finally, since  $J_{p,\epsilon}^{\text{ord}}(\mu) \rightarrow +\infty$  as  $\mu \rightarrow \pm\infty$ , we deduce that the global minimizer of (22) is given by

$$\mu^{(t)} = \min_{i=1, \dots, M} \{J_{p,\epsilon}^{\text{ord}}(r_{(i)})\}. \quad (25)$$

In order to prove the piecewise concavity of  $J_{p,\epsilon}^{\text{ord}}(\mu)$ , for  $\mu \in (r_{(m-1)}, r_{(m)})$ , it suffices to prove that (i)  $g_1(\mu) = ((\mu - r_{(i)})^2 + \frac{\epsilon}{|v_{[i]}|^2})^{p/2}$  is concave for  $i = 1, \dots, m-1$  and (ii)  $g_2(\mu) = ((r_{(i)} - \mu)^2 + \frac{\epsilon}{|v_{[i]}|^2})^{p/2}$  is concave for  $i = m, \dots, M$ . Both functions are twice differentiable, thus it suffices to show that their second derivative is negative. Indeed, for  $g_1(\mu)$  we have that

$$\begin{aligned} g_1''(\mu) &= p \left( (\mu - r_{(i)})^2 + \frac{\epsilon}{|v_{[i]}|^2} \right)^{p/2-1} \\ &\quad \cdot \left( \frac{(p-1)(\mu - r_{(i)})^2 + \frac{\epsilon}{|v_{[i]}|^2}}{(\mu - r_{(i)})^2 + \frac{\epsilon}{|v_{[i]}|^2}} \right). \end{aligned} \quad (26)$$

The first two terms on the right side of (26) are strictly positive, thus we examine the sign of the third term. By noticing that the denominator of this term is positive, we focus on the numerator, which is a second-degree polynomial,

$$\theta(\mu) = (p-1)\mu^2 - 2(p-1)r_{(i)}\mu + \left( (p-1)r_{(i)}^2 + \frac{\epsilon}{|v_{[i]}|^2} \right). \quad (27)$$

The discriminant of the above quadratic equation is equal to  $\Delta = 4(1-p)\frac{\epsilon}{|v_{[i]}|^2}$ . Since  $0 < p < 1$ , we have that  $\Delta > 0$ , thus (27) has two distinct roots,

$$\mu_1 = r_{(i)} - \sqrt{\frac{\epsilon}{|v_{[i]}|^2} \frac{1-p}{1-p}}, \quad \mu_2 = r_{(i)} + \sqrt{\frac{\epsilon}{|v_{[i]}|^2} \frac{1-p}{1-p}}. \quad (28)$$

Then,  $\theta(\mu) < 0$ , and thus  $g_1''(\mu) < 0$ , for  $\mu < \mu_1$  and  $\mu > \mu_2$ . Given that  $i = 1, \dots, m-1$  and  $\mu \in (r_{(m-1)}, r_{(m)})$ , we deduce that  $g_1(\mu)$  is concave if, in addition,  $\mu_2 \leq r_{(m-1)}$ . By solving this inequality in terms of  $\epsilon$ , the concavity of  $g_1(\mu)$  is guaranteed by setting

$$\epsilon \leq \epsilon_{1,m} \triangleq \min_{i=1, \dots, m-1} \{ |v_{[i]}|^2 (1-p)(r_{(m-1)} - r_{(i)})^2 \}. \quad (29)$$

Following a similar procedure for  $g_2(\mu)$ , we end up with the same roots  $\mu_1$  and  $\mu_2$  as in (28). Given that  $i = m, \dots, M$  and  $\mu \in (r_{(m-1)}, r_{(m)}]$ , we deduce that  $g_2(\mu)$  is concave if, in addition,  $r_{(m)} \leq \mu_1$ . By solving this inequality in terms of  $\epsilon$ , the concavity of  $g_2(\mu)$  is guaranteed by setting

$$\epsilon \leq \epsilon_{2,m} \triangleq \min_{i=m, \dots, M} \{ |v_{[i]}|^2 (1-p)(r_{(i)} - r_{(m)})^2 \}. \quad (30)$$

This completes the proof for the piecewise concavity of  $J_{p,\epsilon}^{\text{ord}}(\mu)$  over its domain, by choosing  $\epsilon \leq \min_{m=1, \dots, M} \{\epsilon_{1,m}, \epsilon_{2,m}\}$ . Notice that this requirement gives us also a guideline for setting the value of  $\epsilon$  in the  $\ell_{p,\epsilon}^p$  objective function. ■

We note that the proposed rule for updating the step size, as given by (21), guarantees that the  $\ell_{p,\epsilon}^p$  objective function in (11) does not increase at each iteration. Indeed, the following proposition holds:

*Proposition 2:* Let  $\mathbf{x}_S^{(t+1)} = \mathbf{x}_S^{(t)} + \mu^{(t)*} \mathbf{g}_S^{(t)}$  be the updated sparse solution, where the suboptimal step size  $\mu^{(t)*}$  is calculated by (21). Then, if  $\mathcal{S}^{(t+1)} = \mathcal{S}^{(t)} = \mathcal{S}$ , the proposed update step guarantees the monotonicity of the  $\ell_{p,\epsilon}^p$  objective function, that is,

$$\|\mathbf{y} - \Phi \mathbf{x}_S^{(t+1)}\|_{p,\epsilon}^p \leq \|\mathbf{y} - \Phi \mathbf{x}_S^{(t)}\|_{p,\epsilon}^p, \quad (31)$$

and thus the convergence to zero, due to the nonnegativity of  $\ell_{p,\epsilon}^p$ .

*Proof:* By construction, the suboptimal step size  $\mu^{(t)*}$  calculated by (21) yields an updated sparse solution,  $\mathbf{x}_S^{(t+1)}$ , that minimizes the  $\ell_{p,\epsilon}^p$  of the measurement error, among all  $\mu^{(t)} \in \mathbb{R}$ . From this, we deduce directly that

$$\begin{aligned} \|\mathbf{y} - \Phi \mathbf{x}_S^{(t+1)}\|_{p,\epsilon}^p &= \left\| \mathbf{y} - \Phi \left( \mathbf{x}_S^{(t)} + \mu^{(t)*} \mathbf{g}_S^{(t)} \right) \right\|_{p,\epsilon}^p \\ &\leq \left\| \mathbf{y} - \Phi \left( \mathbf{x}_S^{(t)} + 0 \mathbf{g}_S^{(t)} \right) \right\|_{p,\epsilon}^p = \|\mathbf{y} - \Phi \mathbf{x}_S^{(t)}\|_{p,\epsilon}^p. \end{aligned}$$

Given also that  $\ell_{p,\epsilon}^p$  is lower bounded by zero, we deduce that the proposed update of the sparse solution results in the convergence of the  $\ell_{p,\epsilon}^p$  of the measurement error to zero. ■

If the support of  $\mathbf{x}^{(t+1)}$  differs from the support of  $\mathbf{x}^{(t)}$  estimated at the previous iteration, then the optimality of  $\mu^{(t)}$  may not be guaranteed. To alleviate this issue, a backtracking line search is typically used, that is, if

$$\|\mathbf{y} - \Phi \mathbf{x}^{(t+1)}\|_{p,\epsilon}^p > \|\mathbf{y} - \Phi \mathbf{x}^{(t)}\|_{p,\epsilon}^p,$$

then,  $\mu^{(t)}$  is reduced geometrically,  $\mu^{(t)} \leftarrow c_\mu \cdot \mu^{(t)}$ , where  $c_\mu \in (0, 1)$ , until the objective function in (11) is reduced. Except if mentioned otherwise, in the subsequent evaluations we set the value of the common ratio equal to  $c_\mu = 0.5$ .

Furthermore, in order to improve the convergence performance of MD-IHT, a weighting scheme is employed in our implementation, which assigns small weights to large deviations and large weights to small deviations, as they are computed in the previous iteration. Specifically, the suboptimal step size  $\mu^{(t)}$  is calculated by setting  $\mathbf{u} = (\mathbf{W}^{(t)})^{1/2}(\mathbf{y} - \Phi_S \mathbf{x}_S^{(t)})$  and  $\mathbf{v} = (\mathbf{W}^{(t)})^{1/2} \Phi_S \mathbf{g}_S^{(t)}$  in (22). Doing so, we further suppress the effect of those elements of  $\mathbf{x}^{(t)}$  that yield erroneous contributions to the measurements  $\mathbf{y}$ , while enforcing the contribution

of those elements which better agree with the measurements. We also emphasize that, in contrast to the previous reconstruction techniques, our proposed MD-IHT algorithm does not assume any prior information for the sampling noise, such as the scale of its associated distribution.

It must be noted that, since (13) is a highly nonconvex optimization problem for  $p \in (0, 1)$ , the proposed algorithm converges to a local minimum that depends on the initial point  $\mathbf{x}^{(0)}$ . Nevertheless, given that in practice we are able to achieve accurate reconstruction of the original sparse signal, in conjunction with the theoretical outcomes in [8], [22], [27] that give conditions where the  $\ell_{p,\epsilon}^p$  minimization problem has a unique global minimizer, indicates that the computed local minimizers can be actually very close to the global one. Intuitively, the incorporation of the parameter  $\epsilon$  is critical, since the selection of a relatively large  $\epsilon$  in the weights (16) makes the basin around undesirable local minima more shallow, whilst increasing the probability of making the basin containing the optimal solution deeper. Concerning the selection of  $\mathbf{x}^{(0)}$ , [27] suggests that a good heuristic condition to ensure that the GP updates  $\mathbf{x}^{(t+1)}$  converge to a good approximate global minimizer is to choose  $\mathbf{x}^{(0)}$  such that  $\Phi \mathbf{x}^{(0)} = \mathbf{y}$  and  $\|\mathbf{x}^{(0)}\|_p^p < M/2$ . A solution to this problem can be obtained by applying the primal-dual interior point scheme proposed in [25]. Apart from this initial guess, in our implementation we also support the option of a least-squares initialization, that is,  $\mathbf{x}^{(0)} = (\Phi^T \Phi)^{-1} \Phi^T \mathbf{y}$ . This option is motivated by its successful use in similar  $\ell_p$  ( $0 < p \leq 1$ ) optimization problems (e.g. [9], [28], [32]). Our subsequent experimental evaluation using both synthetic and real data showed that the two initialization schemes perform equally well. If not mentioned otherwise, the first initialization scheme is employed. Nonetheless, the optimal selection of an initial point that guarantees convergence of the regularized  $\ell_{p,\epsilon}^p$  objective to the global minimum is still an open problem.

The algorithm terminates when either a maximum number of iterations,  $\text{maxIter}_{\text{MDIHT}}$ , has been reached, or the relative change of the  $\ell_{p,\epsilon}^p$  objective function is less than a predetermined threshold  $\text{tol}_{\text{MDIHT}}$ . In our implementation we set  $\text{maxIter}_{\text{MDIHT}} = 200$  and  $\text{tol}_{\text{MDIHT}} = 10^{-16}$ . Algorithm 1 summarizes the steps of our MD-IHT sparse reconstruction method.

In the following, we measure the reconstruction quality of the proposed MD-IHT algorithm for the noisy measurements model in (1) by comparing the original sparse signal  $\mathbf{x}_0$  with the reconstructed sparse solution  $\mathbf{x}_0^*$ . Specifically, Theorem 1 below shows that the solution of (11) is an  $s$ -sparse signal with an  $\ell_2$  error that depends on the noise dispersion. This dependence on the  $p$ th FLOM of the noise, instead of its second-order moment that is either infinite or may not exist, yields an increased robustness of MD-IHT to heavy-tailed sampling noise.

*Theorem 1:* Let  $\mathbf{x}_0 \in \mathbb{R}^N$  with  $\mathcal{S} = \text{supp}(\mathbf{x}_0)$  and  $|\mathcal{S}| \leq s$ . Assume  $\Phi \in \mathbb{R}^{M \times N}$  is a measurement matrix with a restricted isometry constant (RIC)  $\delta_{ks} < \sqrt{(k-1)/k}$  for some  $k \geq 4/3$ . Then, for a multivariate additive sampling noise  $\mathbf{n} \in \mathbb{R}^M$  consisting of i.i.d.  $\mathcal{S}\alpha\mathcal{S}$  components with characteristic exponent  $\alpha_n \in [1, 2)$  and dispersion  $\gamma_n \leq \eta$ , the solution of (11),  $\mathbf{x}_0^*$ ,

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**Algorithm 1:** The Proposed MD-IHT Sparse Reconstruction Algorithm.

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**Input:**  $\mathbf{y}$ ,  $\Phi$ ,  $s$ ,  $\epsilon$ ,  $c_\mu$ ,  $\max\text{Iter}_{\text{MDIHT}}$ ,  $\text{tol}_{\text{MDIHT}}$

**Initialize:**

S $\alpha$ S parameters:  $[\alpha_y, \gamma_y] = \text{Mellinfit}(\mathbf{y})$  (\*),  $p = \frac{\alpha_y}{2} - 0.001$

Solution:  $\mathbf{x}^{(0)}$  such that  $\Phi \mathbf{x}^{(0)} = \mathbf{y}$  and  $\|\mathbf{x}^{(0)}\|_p^p < M/2$

Residual:  $\mathbf{r}^{(0)} = \mathbf{y} - \Phi \mathbf{x}^{(0)}$

Weights:  $\mathbf{W}_{i,i}^{(0)} = \frac{1}{\left( (r_i^{(0)})^2 + \epsilon \right)^{1-\frac{p}{2}}}$ ,  $i = 1, \dots, M$  (from (16))

Gradient:  $\mathbf{g}^{(0)} = -p \Phi^T \mathbf{W}^{(0)} \mathbf{r}^{(0)}$  (from (15))

$\text{objVal}_{\text{old}} = \|\mathbf{r}^{(0)}\|_{p,\epsilon}^p$

$\mathcal{S}^{(0)} = \emptyset$ ,  $\text{relChange} = 10^{16}$ ,  $t = 0$

1: **while** ( $\text{relChange} > \text{tol}_{\text{MDIHT}}$  or  $t < \max\text{Iter}_{\text{MDIHT}}$ ) **do**

2: Calculate step size  $\mu^{(t)}$  by solving (22) with  $\mathbf{u} = (\mathbf{W}^{(t)})^{1/2} (\mathbf{y} - \Phi_{\mathcal{S}^{(t)}} \mathbf{x}_{\mathcal{S}^{(t)}}^{(t)})$  and

$\mathbf{v} = (\mathbf{W}^{(t)})^{1/2} \Phi_{\mathcal{S}^{(t)}} \mathbf{g}_{\mathcal{S}^{(t)}}^{(t)}$

3: Update solution via (14):  $\mathbf{x}^{(t+1)} = \mathbf{H}_s(\mathbf{x}^{(t)} + \mu^{(t)} \mathbf{g}^{(t)})$

4: Update support  $\mathcal{S}^{(t+1)} = \{j \mid x_j^{(t+1)} \neq 0\}$

5: Update residual  $\mathbf{r}^{(t+1)} = \mathbf{y} - \Phi_{\mathcal{S}^{(t+1)}} \mathbf{x}_{\mathcal{S}^{(t+1)}}^{(t+1)}$

6:  $\text{objVal}_{\text{new}} = \|\mathbf{r}^{(t+1)}\|_{p,\epsilon}^p$   
# Perform backtracking if necessary #

7: **if**  $\mathcal{S}^{(t+1)} \neq \mathcal{S}^{(t)}$  **then**

8: **while**  $\text{objVal}_{\text{new}} > \text{objVal}_{\text{old}}$  **do**

9:  $\mu^{(t)} \leftarrow c_\mu \cdot \mu^{(t)}$

10: Update solution via (14):

$\mathbf{x}^{(t+1)} = \mathbf{H}_s(\mathbf{x}^{(t)} + \mu^{(t)} \mathbf{g}^{(t)})$

11: Update support  $\mathcal{S}^{(t+1)} = \{j \mid x_j^{(t+1)} \neq 0\}$

12: Update residual  $\mathbf{r}^{(t+1)} = \mathbf{y} - \Phi_{\mathcal{S}^{(t+1)}} \mathbf{x}_{\mathcal{S}^{(t+1)}}^{(t+1)}$

13:  $\text{objVal}_{\text{new}} = \|\mathbf{r}^{(t+1)}\|_{p,\epsilon}^p$

14: **end while**

15: **end if**

16:  $\text{relChange} = |\text{objVal}_{\text{new}} - \text{objVal}_{\text{old}}| / \text{objVal}_{\text{new}}$

17:  $\text{objVal}_{\text{old}} = \text{objVal}_{\text{new}}$

18: Update weights  $\mathbf{W}^{(t+1)}$  (from (16))

19: Update gradient  $\mathbf{g}^{(t+1)} = -p \Phi^T \mathbf{W}^{(t+1)} \mathbf{r}^{(t+1)}$  (from (15))

20:  $t = t + 1$

21: **end while**

**Output:** The original sparse signal  $\mathbf{x}_0^* = \mathbf{x}^{(t)}$

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(\*) Mellinfit( $\mathbf{y}$ ) denotes the algorithm described in Section III-C for the estimation of the S $\alpha$ S model parameters from the noisy measurements  $\mathbf{y}$  using the method of log-cumulants.

**Note:** In the general case, the measurement operator  $\Phi$  is replaced by  $\mathbf{A} = \Phi \Psi^T$ . Then, MD-IHT estimates a sparse coefficients vector  $\alpha_0^*$  and the original signal is obtained by  $\mathbf{x}_0^* = \Psi^T \alpha_0^*$ .

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yields a bounded reconstruction error,

$$\|\mathbf{x}_0 - \mathbf{x}_0^*\|_2 \leq C_0 C_{p,\alpha_n} \eta, \quad (32)$$

where the constant  $C_0$  depends on  $k$ ,  $\delta_{ks}$ ,  $M$  and  $p$ , and  $C_{p,\alpha_n}$  is given by (7).

*Proof:* Following a similar approach as in [30], we set  $\mathbf{x}_0^* = \mathbf{x}_0 + \mathbf{h}$ , where  $\mathbf{h}$  is a perturbation of the original sparse signal  $\mathbf{x}_0$ . Since  $\mathbf{x}_0^*$  is a feasible point and the dispersion of the error (i.e., the noise) is assumed to be bounded,  $\gamma_n \leq \eta$ , it follows that

$$\begin{aligned} \|\Phi \mathbf{h}\|_2 &= \|\Phi \mathbf{x}_0^* - \Phi \mathbf{x}_0\|_2 = \|(\Phi \mathbf{x}_0^* - \mathbf{y}) + (\mathbf{y} - \Phi \mathbf{x}_0)\|_2 \\ &\stackrel{(a)}{\leq} \|\Phi \mathbf{x}_0^* - \mathbf{y}\|_2 + \|\mathbf{y} - \Phi \mathbf{x}_0\|_2 \stackrel{(b)}{\leq} \|\Phi \mathbf{x}_0^* - \mathbf{y}\|_p + \|\mathbf{y} - \Phi \mathbf{x}_0\|_p \\ &\leq 2 \|\mathbf{n}\|_p = 2 \left( \sum_{j=1}^M \mathbb{E}\{|n_j|^p\} \right)^{1/p} \stackrel{(c)}{=} 2 \left( \sum_{j=1}^M \gamma_n^p \mathbb{E}\{|Z_j|^p\} \right)^{1/p} \\ &\stackrel{(d)}{=} 2 M^{1/p} \gamma_n C_{p,\alpha_n} \stackrel{\gamma_n \leq \eta}{\leq} 2 M^{1/p} (C_{p,\alpha_n} \eta), \end{aligned} \quad (33)$$

where (a) follows from the triangle inequality and (b) from the property that if  $0 < p < q$  then  $\|\mathbf{x}\|_q \leq \|\mathbf{x}\|_p$  (in our case  $0 < p < 1$ ). The equality (c) is obtained by writing  $n_j = \gamma_n Z_j$ ,  $j = 1, \dots, M$ , where  $Z_j$  are standardized S $\alpha$ S random variables, i.e.,  $Z_j \sim f_{\alpha_n}(1, 0)$ , and then combining with (6) to get  $\mathbb{E}\{|Z_j|^p\} = (C_{p,\alpha_n})^p$ , and subsequently the equality (d) (ref. [14], pg. 18).

To complete the proof, we utilize an intermediate result shown in [31] (see proof of Theorem 2.1), which states that if  $\delta_{ks} < \sqrt{(k-1)/k}$  for some  $k \geq 4/3$  then, we have

$$\|\mathbf{h}\|_2 \leq \frac{\sqrt{2(1+\delta_{ks})} \|\Phi \mathbf{h}\|_2}{1 - \sqrt{k/(k-1)} \delta_{ks}}. \quad (34)$$

By combining (33) and (34) we have

$$\|\mathbf{h}\|_2 \leq \frac{2 M^{1/p} \sqrt{2(1+\delta_{ks})}}{1 - \sqrt{k/(k-1)} \delta_{ks}} (C_{p,\alpha_n} \eta), \quad (35)$$

which is the desired result for  $C_0 = \frac{2 M^{1/p} \sqrt{2(1+\delta_{ks})}}{1 - \sqrt{k/(k-1)} \delta_{ks}}$ . ■

From (32) we deduce that as the noise dispersion  $\gamma_n \rightarrow 0$  the reconstruction error approaches zero, whereas in the noiseless case ( $\eta = 0$ ) the reconstruction is perfect. Furthermore, the factor  $M^{1/p}$  in (35) expresses the dependence of the reconstruction error on both the size and impulsiveness of the noise vector. In addition, the above RIC condition,  $\delta_{ks} < \sqrt{(k-1)/k}$ , proposed by [31], yields sharp restricted isometry property (RIP) conditions on the higher order RICs, which can be satisfied by a significantly larger set of random measurement matrices in some settings.

#### IV. PERFORMANCE EVALUATION

This section evaluates the efficiency of our proposed MD-IHT algorithm as a robust sparse reconstruction method under impulsive sampling noise. To this end, numerical experiments are performed with synthetic signals, as well as with real data,

along with a comparison against state-of-the-art sparse reconstruction methods tailored to impulsive noise. In particular, the following methods are used for comparisons, which recover the original sparse signal by solving either  $\ell_2$  or  $\ell_p$  ( $p \leq 2$ ) optimization problems: 1) orthogonal matching pursuit (OMP) [1],<sup>3</sup> 2)  $\ell_p$ -reweighted least squares (LpRLS) [32],<sup>4</sup> and 3) Lorentzian iterative hard thresholding (LIHT) [33]. Appropriate parameters tuning is done for each algorithm according to the guidelines of the associated toolboxes, in order to achieve the optimal reconstruction performance. For the OMP we assume that the noise tolerance  $\eta$  is known and used as a stopping criterion.

### A. Experiments With Synthetic Signals

The synthetic signals are generated using the following settings, unless stated otherwise: signal length  $N = 1024$ ; cardinality of the sparse support  $s = |\mathcal{S}| = \lceil 2\%N \rceil$ ; the nonzero coefficients are drawn from a Student's- $t$  distribution with one degree of freedom and their positions are chosen uniformly at random from the index set  $\{1, 2, \dots, N\}$ ; the DCT matrix is used as the sparsifying dictionary  $\Psi$ , that is, the measurement operator is given by  $\mathbf{A} = \Phi\Psi^T$ ; the measurement matrix  $\Phi$  has i.i.d. entries drawn from a Bernoulli distribution  $\{-1, +1\}$  with equal probability; the number of linear random measurements is set to  $M = \lceil 25\%N \rceil$  unless otherwise specified. Furthermore, the results of each experiment are averaged over 500 Monte Carlo repetitions with different realizations of the sparse signals, the random measurement matrices, and the additive noise term. The reconstruction quality is measured in terms of the signal-to-error ratio (SER) (in dB) defined by

$$\text{SER}(\mathbf{x}, \hat{\mathbf{x}}) = 10 \log_{10} \left( \frac{\sum_{j=1}^N x_j^2}{\sum_{j=1}^N (x_j - \hat{x}_j)^2} \right), \quad (36)$$

where  $\mathbf{x}$  and  $\hat{\mathbf{x}}$  correspond to the original and reconstructed signals, respectively.

First, we examine the performance of MD-IHT as a function of the noise strength. To this end, we vary the noise dispersion  $\gamma_n \in \{0.001, 0.01, 0.1, 1\}$ , whilst fixing the noise characteristic exponent  $\alpha_n \in \{1, 1.5\}$ . The choice of  $\alpha_n = 1$  (i.e., Cauchy distribution) is made for a fair comparison with LIHT which is derived from Cauchy statistics. The number of linear projections is set to  $M = \lceil 25\%N \rceil$  and the noise tolerance  $\eta = M\gamma_n^2$  (for the OMP). Furthermore, all the required S $\alpha$ S parameters in Algorithm 1 are estimated from the noisy measurements directly. Fig. 1 shows the reconstruction performance of each method, in terms of the achieved SER averaged over 500 Monte Carlo runs. As the noise strength increases, the reconstruction accuracy of all methods decreases, as expected. However, MD-IHT yields the highest accuracy among the four methods for the whole range of  $\gamma_n$  and for both the  $\alpha_n$  values. Especially in the Cauchy case ( $\alpha_n = 1$ ), this reveals that our MD-IHT algorithm outperforms LIHT, which is tailored to Cauchy statistics, thus demonstrating an increased robustness of MD-IHT to a broader

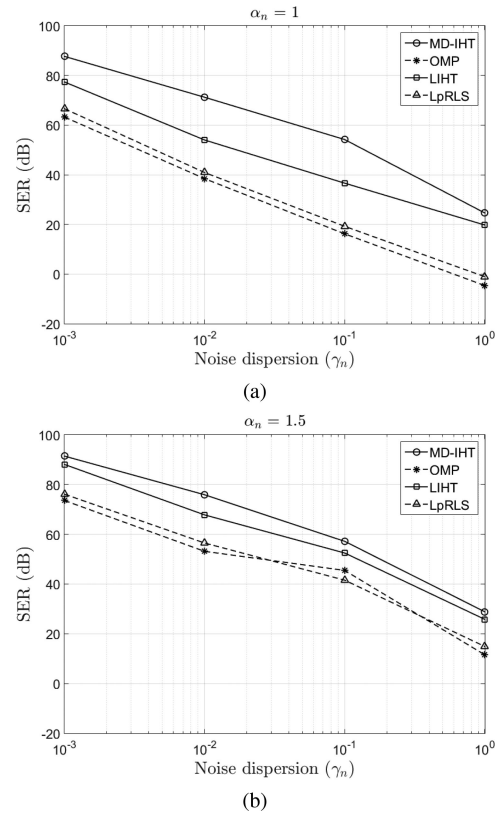


Fig. 1. Comparison of reconstruction error as a function of noise strength for MD-IHT, OMP, LpRLS, and LIHT. Linear projections are used corrupted by S $\alpha$ S sampling noise with  $\alpha_n \in \{1, 1.5\}$  and  $\gamma_n \in \{0.001, 0.01, 0.1, 1\}$ . Average SER is shown over 500 Monte Carlo runs.

range of noise strength and impulsiveness. Indeed, in contrast to the LIHT, whose performance is controlled by tuning only the scale parameter of the Lorentzian norm, the performance of MD-IHT depends on both the value of  $p$  in the  $\ell_{p,\epsilon}^p$  optimization and the estimated dispersion from the noisy measurements. Furthermore, OMP results in the worst reconstruction quality, illustrating the inefficiency of  $\ell_2$ -based methods to suppress the presence of infinite variance sampling noise in the random measurements.

As a second experiment, we evaluate the performance of MD-IHT as a function of the noise impulsiveness. To this end, we vary the noise characteristic exponent  $\alpha_n \in [1, 2)$ , whilst fixing the noise dispersion  $\gamma_n \in \{0.01, 0.1\}$ . As before, the number of linear projections is set to  $M = \lceil 25\%N \rceil$  and the noise tolerance  $\eta = M\gamma_n^2$  (for the OMP). Fig. 2 shows the reconstruction performance of each method, in terms of the achieved SER averaged over 500 Monte Carlo runs. As the noise impulsiveness decreases (i.e.,  $\alpha_n \rightarrow 2$ ) the reconstruction accuracy of all methods increases, whilst they all result in a comparable average SER. As expected, the lower the noise dispersion,  $\gamma_n$ , the higher the reconstruction quality for the four methods. Furthermore, MD-IHT outperforms the other three methods over the whole range of  $\alpha_n$  and for both noise dispersion values. Interestingly, both the MD-IHT and the LIHT algorithms, which yield the same performance for Gaussian sampling noise,

<sup>3</sup>MATLAB code available from <https://goo.gl/VHvyJe>.

<sup>4</sup>MATLAB code available from <https://sites.google.com/site/igorcarron2/cscodes>.

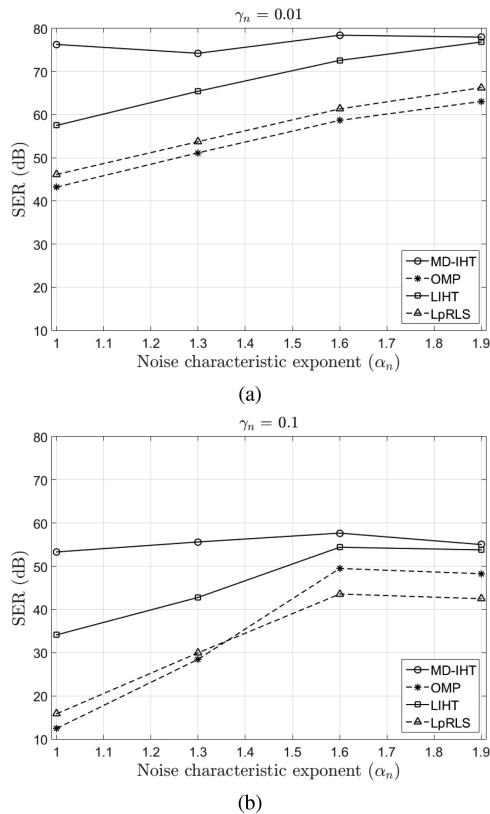


Fig. 2. Comparison of reconstruction error as a function of noise impulsiveness for MD-IHT, OMP, LpRLS, and LIHT. Linear projections are used corrupted by  $S\alpha S$  sampling noise with  $\alpha_n \in [1, 2)$  and  $\gamma_n \in \{0.01, 0.1\}$ . Average SER is shown over 500 Monte Carlo runs.

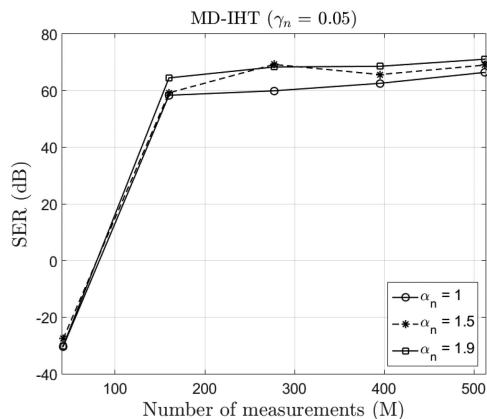


Fig. 3. Average SER of MD-IHT as a function of the number of linear measurements corrupted by  $S\alpha S$  sampling noise with  $\alpha_n \in \{1, 1.5, 1.9\}$  and  $\gamma_n = 0.05$ .

outperform clearly the least squares-based OMP method, which better adapts to light-tailed environments.

As a last experiment, we examine the reconstruction performance of MD-IHT as the number of linear measurements,  $M$ , varies from  $2s$  (i.e., twice the cardinality of the sparse support) to  $N/2$ , for a varying sampling noise impulsiveness with  $\alpha_n \in \{1, 1.5, 1.9\}$ , and a fixed  $\gamma_n = 0.05$ . Fig. 3 shows that MD-IHT starts yielding fair reconstructions of the original signals using only  $M \approx 10\%N$  corrupted linear measurements.

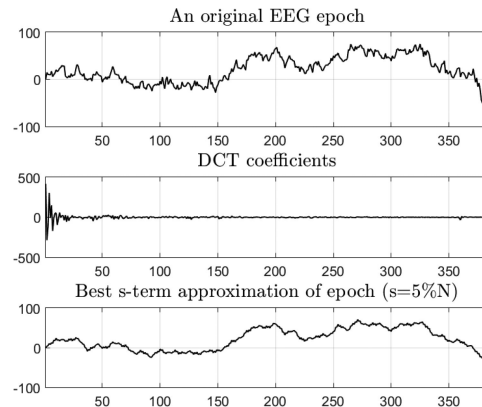


Fig. 4. Original EEG epoch, DCT coefficients and best  $s$ -term approximation ( $s = \lceil 5\%N \rceil$ ).

Most importantly, this observation holds even for heavily corrupted measurements (i.e., small  $\alpha_n$  values), which illustrates the robustness of MD-IHT in a broad range of impulsive environments. However, as the noise impulsiveness increases, more measurements are required to achieve a satisfactory reconstruction quality. This is an expected result, which also resembles the conventional  $\ell_2$ -based reconstruction methods that require more measurements as the noise variance increases.

### B. Experiments with EEG Data

The following experiments evaluate and compare the reconstruction performance of our proposed algorithm on real data [34]. Specifically, the utilized dataset contains electroencephalography (EEG) signals of 32 channels with sequence length of 30720 data points. Each channel signal consists of 80 epochs, each one containing  $N = 384$  points. Artifacts caused by muscle movement also occur in the signals. The EEG signals are compressively sampled in an epoch-by-epoch fashion using a Bernoulli matrix  $\Phi$ , whereas the  $384 \times 384$  DCT matrix is used as the sparsifying dictionary  $\Psi$ . The sparsity level is fixed at  $s = \lceil 5\%N \rceil$ , whilst the number of linear measurements is set to  $M = N/2$ . The measurements are contaminated by additive  $S\alpha S$  noise with  $\alpha_n \in \{1, 1.5\}$  and  $\gamma_n \in \{0.1, 0.5, 1.5\}$ . The reconstruction quality is measured in terms of the structural similarity index (SSIM) for 1D signals [35], which is a better performance index for structured signals. The higher the SSIM value the better the reconstruction quality, with a value equal to 1 corresponding to an excellent reconstruction. Furthermore, the length of the sliding window for the calculation of SSIM is set to 100.

Fig. 4 shows an original EEG epoch, along with the corresponding DCT coefficients and the best  $s$ -term approximation ( $s = \lceil 5\%N \rceil$ ). Clearly, the signal is nonsparse both in time and frequency. Nevertheless, for the reconstruction we consider that the targeted sparsity level of the DCT coefficients is upper bounded by  $s = \lceil 5\%N \rceil$ . In Fig. 5, the corresponding reconstructed epochs are plotted, by applying MD-IHT, OMP, LIHT and LpRLS on linear measurements corrupted by heavy-tailed noise with parameters  $\alpha_n = 1, \gamma_n = 1.5$ . For this specific epoch, MD-IHT and LIHT yield the best reconstruction (i.e., closest to

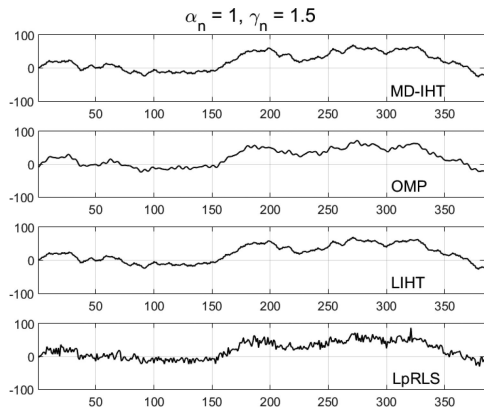


Fig. 5. Reconstructed EEG epoch from noisy measurements ( $\alpha_n = 1$ ,  $\gamma_n = 1.5$ ) using MD-IHT, OMP, LIHT, and LpRLS.

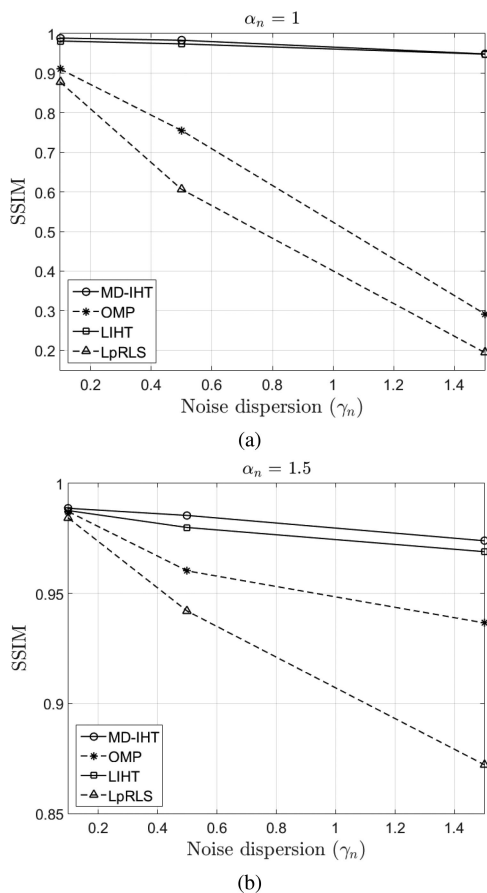


Fig. 6. Comparison of reconstruction SSIM as a function of noise dispersion for MD-IHT, OMP, LIHT, and LpRLS. Linear projections are used corrupted by  $S\alpha S$  sampling noise with (a)  $\alpha_n = 1$  and (b)  $\alpha_n = 1.5$ . The SSIM values are averaged over all the epochs and Monte Carlo runs.

the best  $s$ -term approximation), followed by OMP and LpRLS, which result in a smoother and more oscillatory, respectively, reconstructed epoch.

In order to study the effects of various impulsive behaviors for the corrupting noise, we vary  $\alpha_n$  in  $\{1, 1.5\}$  and  $\gamma_n$  in  $\{0.1, 0.5, 1.5\}$ . The choice of  $\alpha_n = 1$  is made for a fair comparison with LIHT, which is best adapted to Cauchy statistics for the sampling noise. Figs. 6a–6b show the reconstruction

performance, in terms of the SSIM averaged over all the epochs and Monte Carlo runs, for MD-IHT, OMP, LIHT, and LpRLS. The linear measurements are corrupted by  $S\alpha S$  sampling noise with  $\alpha_n = 1$  and  $\alpha_n = 1.5$ , respectively. Clearly, as the noise impulsiveness increases ( $\alpha_n = 1$ ), OMP and LpRLS fail to achieve a fair reconstruction of the epochs for increasing noise dispersion. On the contrary, MD-IHT and LIHT are highly robust over the whole range of  $\gamma_n$  values, with MD-IHT presenting a slightly better reconstruction when compared with LIHT. This is very important if we notice that LIHT is intrinsically related with a Cauchy model for the sampling noise. A similar behavior is observed when  $\alpha_n = 1.5$ . In particular, the performance of OMP and LpRLS improves as the noise impulsiveness decreases. However, both methods are still inferior against MD-IHT and LIHT, which better adapt to heavy-tailed environments. As before, MD-IHT evidently achieves a more accurate reconstruction, on average, when compared with LIHT for all the dispersion values.

## V. CONCLUSIONS AND FUTURE WORK

In this paper, a robust method was proposed for the reconstruction of sparse signals whose compressive measurements are corrupted by impulsive sampling noise. More specifically, the heavy-tailed statistics of sampling noise, with possibly infinite variance, was modeled by means of  $S\alpha S$  distributions. Subsequently, the effects of additive impulsive sampling noise were suppressed by designing a novel iterative hard thresholding method based on a minimum dispersion (MD) optimization criterion. This criterion emerges naturally in the case of additive sampling noise modeled by  $S\alpha S$  distributions. The proposed MD-IHT algorithm demonstrated an increased robustness against gross outliers through a least  $\ell_{p,\epsilon}^p$  estimation error criterion, where  $p$  depends on the inherent impulsiveness of the noise. A reconstruction error bound was derived that depends on the noise strength, along with rules for tuning the key parameters, such as the value of  $p$  and the gradient projection step size, in order to guarantee convergence for a broad range of impulsive noise behaviors. Experimental evaluations with synthetic and real data revealed that MD-IHT outperforms significantly state-of-the-art methods in the case of highly impulsive sampling noise, whilst resulting in a comparable performance in light-tailed environments.

However, a theoretical framework for selecting the optimal values of the key parameters for the MD-IHT algorithm is still an open question. Given the importance of initializing appropriately our proposed iterative algorithm to guarantee convergence to the global minimum, using convex results to initialize an efficient search for a locally optimal nonconvex solution could combine the strengths of convex and nonconvex formulations. Furthermore, incorporating some prior knowledge about the unknown sparse support in the reconstruction process typically improves the reconstruction quality. To address this issue, we will examine a modification of the MD-IHT algorithm for stable recovery from compressive measurements given a partially known support. Finally, our current study focused on multivariate sampling noise with i.i.d. components. It would be also of

interest to investigate the more generic case of sampling noise with dependent, jointly  $S\alpha S$  components.

APPENDIX  
PROOF OF POSITIVITY OF  $C_{p,\alpha}^{-p}$

Given the definition in (7), in order to prove the positivity of  $C_{p,\alpha}^{-p}$  in (10), it suffices to show that

$$\frac{\Gamma\left(1 - \frac{p}{\alpha}\right)}{\cos\left(\frac{\pi}{2}p\right)\Gamma(1-p)} > 0. \quad (37)$$

Indeed, the requirement for the existence of FLOMs induces that  $0 < p < \alpha$ . This, combined with our empirical rule for setting the optimal value of  $p$  as  $p \lesssim \frac{\alpha}{2}$  (ref. Section II), and our focus on  $\alpha \in [1, 2)$ , yield  $0 < p < 1$ . Based on this, the sign of each term in (37) is as follows:

- i)  $p \lesssim \frac{\alpha}{2} \Rightarrow 1 - \frac{p}{\alpha} \gtrsim \frac{1}{2} > 0 \Rightarrow \Gamma\left(1 - \frac{p}{\alpha}\right) > 0$
- ii)  $0 < p < 1 \Rightarrow 0 < \frac{\pi}{2}p < \frac{\pi}{2} \Rightarrow 1 > \cos\left(\frac{\pi}{2}p\right) > 0$
- iii)  $0 < p < 1 \Rightarrow 1 > 1 - p > 0 \Rightarrow \Gamma(1-p) > 0$

From (i)–(iii) we deduce that all the terms are positive, which proves that  $C_{p,\alpha}^{-p} > 0$  for  $\alpha \in [1, 2)$ . Notice though that, in fact, the above result holds for the whole range of feasible values of  $\alpha \in (0, 2]$ .

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REFERENCES

- [1] J. Tropp and A. Gilbert, "Signal recovery from random measurements via orthogonal matching pursuit," *IEEE Trans. Inf. Theory*, vol. 53, no. 12, pp. 4655–4666, Dec. 2007.
- [2] W. Dai and O. Milenkovic, "Subspace pursuit for compressive sensing signal reconstruction," *IEEE Trans. Inf. Theory*, vol. 55, no. 5, pp. 2230–2249, May 2009.
- [3] D. Donoho, M. Elad, and V. Temlyakov, "Stable recovery of sparse overcomplete representations in the presence of noise," *IEEE Trans. Inf. Theory*, vol. 52, no. 1, pp. 6–18, Jan. 2006.
- [4] J. Tropp, "Just relax: Convex programming methods for identifying sparse signals in noise," *IEEE Trans. Inf. Theory*, vol. 52, no. 3, pp. 1030–1051, Mar. 2006.
- [5] M. Figueiredo, R. Nowak, and S. Wright, "Gradient projection for sparse reconstruction: Application to compressed sensing and other inverse problems," *IEEE J. Sel. Topics Signal Process.*, vol. 1, no. 4, pp. 586–597, Dec. 2007.
- [6] S. Babacan, R. Molina, and A. Katsaggelos, "Bayesian compressive sensing using Laplace priors," *IEEE Trans. Image Process.*, vol. 19, no. 1, pp. 53–63, Jan. 2010.
- [7] X. Tan and J. Li, "Computationally efficient sparse Bayesian learning via belief propagation," *IEEE Trans. Signal Process.*, vol. 58, no. 4, pp. 2010–2021, Apr. 2010.
- [8] R. Chartrand, "Exact reconstruction of sparse signals via nonconvex minimization," *IEEE Signal Process. Lett.*, vol. 14, no. 10, pp. 707–710, Oct. 2007.
- [9] E. Candès, M. Wakin, and S. Boyd, "Enhancing sparsity by reweighted  $\ell_1$  minimization," *J. Fourier Anal. Appl.*, vol. 14, no. 5, pp. 877–905, Dec. 2008.
- [10] F. Liu *et al.*, "Nonconvex compressed sensing by nature-inspired optimization algorithms," *IEEE Trans. Cybern.*, vol. 45, no. 5, pp. 1042–1053, May 2015.
- [11] A. Miller, *Subset Selection in Regression*, 2nd ed. London, U.K.: Chapman and Hall/CRC, 2002.
- [12] A. B. Ramirez, G. R. Arce, D. Otero, J.-L. Paredes, and B. M. Sadler, "Reconstruction of sparse signals from  $\ell_1$  dimensionality-reduced Cauchy random projections," *IEEE Trans. Signal Process.*, vol. 60, no. 11, pp. 5725–5737, Nov. 2012.
- [13] R. Carrillo, K. Barner, and T. Aysal, "Robust sampling and reconstruction methods for sparse signals in the presence of impulsive noise," *IEEE J. Sel. Topics Signal Process.*, vol. 4, no. 2, pp. 392–408, Apr. 2010.
- [14] G. Samorodnitsky and M. Taqqu, *Stable Non-Gaussian Random Processes: Stochastic Models With Infinite Variance*. New York, NY, USA: Chapman & Hall, 1994.
- [15] J. P. Nolan, "Numerical calculation of stable densities and distribution functions," *Commun. Statist. Stoch. Models*, vol. 13, no. 4, pp. 759–774, 1997.
- [16] E. Candès and P. Randall, "Highly robust error correction by convex programming," *IEEE Trans. Inf. Theory*, vol. 54, no. 7, pp. 2829–2840, Jul. 2008.
- [17] B. Popilka, S. Setzer, and G. Steidl, "Signal recovery from incomplete measurements in the presence of outliers," *Inverse Problems Imag.*, vol. 1, no. 4, pp. 661–672, Nov. 2007.
- [18] J. Laska, M. Davenport, and R. Baraniuk, "Exact signal recovery from sparsely corrupted measurements through the pursuit of justice," in *Proc. 43rd Asilomar Conf. Signals, Syst. Comput.*, Pacific Grove, CA, USA, Nov. 2009, pp. 1556–1560.
- [19] A. Ramirez, G. R. Arce, and B. Sadler, "Fast algorithms for reconstruction of sparse signals from Cauchy random projections," in *Proc. 18th Eur. Signal Process. Conf.*, Aalborg, Denmark, Aug. 2010, pp. 432–436.
- [20] G. Tzagkarakis, B. Beferull-Lozano, and P. Tsakalides, "Rotation-invariant texture retrieval with Gaussianized steerable pyramids," *IEEE Trans. Image Process.*, vol. 15, no. 9, pp. 2702–2718, Sep. 2006.
- [21] C. Nikias and M. Shao, *Signal Processing with Alpha-Stable Distributions and Applications*. Hoboken, NJ, USA: Wiley, 1995.
- [22] R. Chartrand and V. Staneva, "Restricted isometry properties and nonconvex compressive sensing," *Inverse Problems*, vol. 24, no. 3, 2008, Art. no. 035020.
- [23] D. Ge, X. Jiang, and Y. Ye, "A note on the complexity of  $L_p$  minimization," *Math. Program.*, vol. 129, no. 2, pp. 285–299, Oct. 2011.
- [24] T. Blumensath and M. Davies, "Iterative hard thresholding for compressed sensing," *Appl. Comput. Harmon. Anal.*, vol. 27, no. 3, pp. 265–274, Nov. 2009.
- [25] S. Foucart and M. J. Lai, "Sparsest solutions of underdetermined linear systems via  $\ell_q$  minimization for  $0 < q \leq 1$ ," *Appl. Comput. Harmon. Anal.*, vol. 26, no. 3, pp. 395–407, 2009.
- [26] S. Foucart and M. J. Lai, "Sparse recovery with pre-Gaussian random matrices," *Studia Math.*, vol. 200, pp. 91–102, 2010.
- [27] M. J. Lai and J. Wang, "An unconstrained  $\ell_q$  minimization with  $0 < q \leq 1$  for sparse solution of underdetermined linear systems," *SIAM J. Optim.*, vol. 21, no. 1, pp. 82–101, 2011.
- [28] A. Achim *et al.*, "Compressive sensing for ultrasound RF echoes using alpha-stable distributions," in *Proc. 32nd Annu. Int. Conf. Med. Biol. Soc.*, Buenos Aires, Argentina, Sep. 2010, pp. 4304–4307.
- [29] Y. Li and G. Arce, "A maximum likelihood approach to least absolute deviation regression," *EURASIP J. Appl. Signal Process.*, vol. 12, pp. 1762–1769, 2004.
- [30] E. Candès, "The restricted isometry property and its implications for compressed sensing," *Comptes Rendus Mathématique*, vol. 346, pp. 589–592, 2008.
- [31] T. Cai and A. Zhang, "Sparse representation of a polytope and recovery of sparse signals and low-rank matrices," *IEEE Trans. Inf. Theory*, vol. 60, no. 1, pp. 122–132, Jan. 2014.
- [32] R. Chartrand and W. Yin, "Iteratively reweighted algorithms for compressive sensing," in *Proc. IEEE Int. Conf. Acoust., Speech, Signal Process.*, Las Vegas, NV, USA, Mar./Apr. 2008, pp. 3869–3872.
- [33] R. Carrillo and K. Barner, "Lorentzian iterative hard thresholding: Robust compressed sensing with prior information," *IEEE Trans. Signal Process.*, vol. 61, no. 19, pp. 4822–4833, Oct. 2013.
- [34] A. Delorme and S. Makeig, "EEGLAB: An open source toolbox for analysis of single-trial EEG dynamics including independent component analysis," *J. Neurosci. Methods*, vol. 134, no. 1, pp. 9–21, 2004.
- [35] Z. Wang and A. Bovik, "Mean squared error: Love it or leave it? A new look at signal fidelity measures," *IEEE Signal Process. Mag.*, vol. 26, no. 1, pp. 98–117, Jan. 2009.



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